CA1391, CA1394

OBSOLETE PRODUCT
NO RECOMMENDED REPLACEMENT
Call Central Applications 1-800-442-7747
or email: centapp@harris.com

May 1999

TV Horizontal Processors

Features

- **CA1391E - Positive Horizontal Sawtooth Input**
- **CA1394E - Negative Horizontal Sawtooth Input**
- **Internal Shunt Regulator**
- **Linear Balanced Phase Detector**
- **Preset Hold Control Capability**
- **Pull-In** . . . . . ±**300Hz (Typ)**
- **Low Thermal Frequency Drift**
- **Small Static Phase Error**
- **Variable Output Duty Cycle**
- **Adjustable DC Loop Gain**

Description

The Harris CA1391E and CA1394E are monolithic integrated circuits designed for use in the low-level horizontal section of monochrome or color television receivers. Functions include a phase detector, an oscillator, a regulator, and a pre-driver.

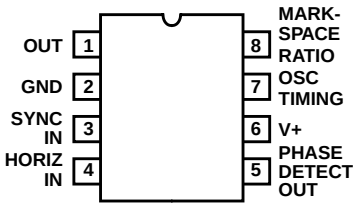
The CA1391E and CA1394E are electrically equivalent and pin compatible with industry types 1391 and 1394 in similar packages.

Part Number Information

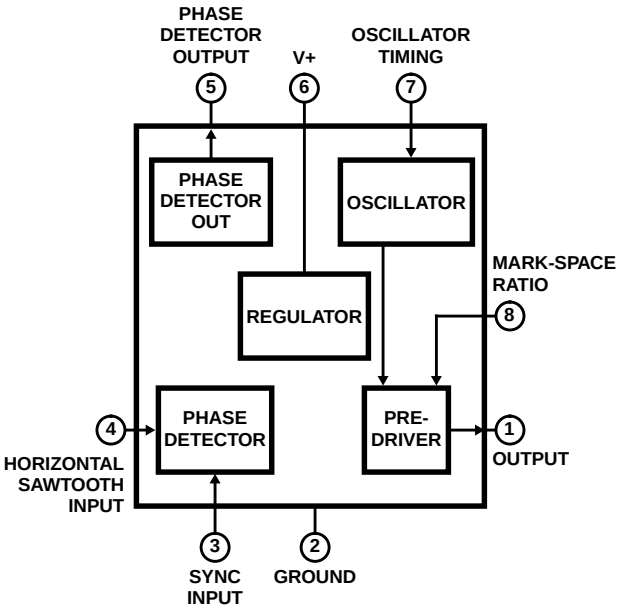
PART NUMBER	TEMP. RANGE (^{O}C)	PACKAGE	PKG. NO.
CA1391E	0 to 85	8 Ld PDIP	E8.3
CA1394E	0 to 85	8 Ld PDIP	E8.3

Pinout

CA1391, CA1394
(PDIP)
TOP VIEW



Functional Diagram



CAUTION: These devices are sensitive to electrostatic discharge. Users should follow proper IC Handling Procedures.

File Number **981.4**

Absolute Maximum Ratings

DC Supply Current 40mA
DC Output Voltage 40V
DC Output Current 30mA
Sync Input Voltage . . . . . $5V_{P-P}$
Sawtooth Input Voltage . . . . . $5V_{P-P}$

Thermal Information

Thermal Resistance (Typical, Note 1) θ_{JA} ($^oC/W$)
PDIP Package 120
Maximum Junction Temperature (Plastic Package) . . . . . 150^oC
Maximum Storage Temperature Range . . . . . -65^oC to 150^oC
Maximum Lead Temperature (Soldering 10s) . . . . . 300^oC

Operating Conditions

Temperature Range . . . . . 0^oC to 85^oC

CAUTION: Stresses above those listed in "Absolute Maximum Ra ings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

NOTE:

1. θ_{JA} is measured with the component mounted on an evaluation PC board in free air.

Electrical Specifications (See Figure 1)

PARAMETER	TEST CONDITIONS	TEMP. (^oC)	MIN	TYP	MAX	UNITS
Supply Voltage	S_1, S_5, S_6 = 2; S_2, S_3, S_4, S_7, S_8 = 1 Measure Terminal 6 to GND	25	8	-	9	V
Free Running Frequency -1%	S_1, S_5, S_6 = 2; S_2, S_3, S_4, S_7, S_8 = 1 Counter to Terminal 1	25	14734	-	16734	Hz
Output Leakage	S_2, S_3, S_6, S_8 = 1; S_1, S_4, S_5, S_7 = 2 Measure Terminal 1 to 25V	25	-	10	-	mV
Output Saturation	S_2, S_3, S_5, S_6, S_8 = 1; S_1, S_4, S_7 = 2 Measure Terminal 1 to GND	25	-	60	-	mV
Phase Detector Bias	S_2, S_5, S_6, S_8 = 1; S_1, S_3, S_4, S_7 = 2 Measure Terminal 3 to GND	25	-	1.9	-	V
Phase Detector Leak	S_5, S_8 = 1; S_1, S_2, S_3, S_4, S_6, S_7 = 2 Measure Terminal 5 to +4V	25	-2	-	2	mV
Phase Detector Low	S_1, S_5, S_8 = 1; S_2, S_3, S_4, S_6, S_7 = 2 Measure Terminal 5 to +4V	25	-0.55 (Note 2)	-	-	V
Phase Detector High	S_1, S_5, S_6, S_8 = 1; S_2, S_3, S_4, S_7 = 2 Measure Terminal 5 to +4V	25	+0.55 (Note 2)	-	-	V
Phase Detector Balance	$V_{DET2} + V_{DET3}$	25	-100	-	100	mV
Sync Diode	S_1, S_2, S_3, S_4, S_6, S_7 = 1; S_5, S_8 = 2	25	0.3	-	1.2	V
Static Phase Error	See Figure 3	25	-	0.5	-	µs
Oscillator Pull In Range			-	±300	-	Hz
Oscillator Hold In Range			-	±900	-	Hz

NOTE:

2. Polarity reversed in the CA1391.

Test Circuit

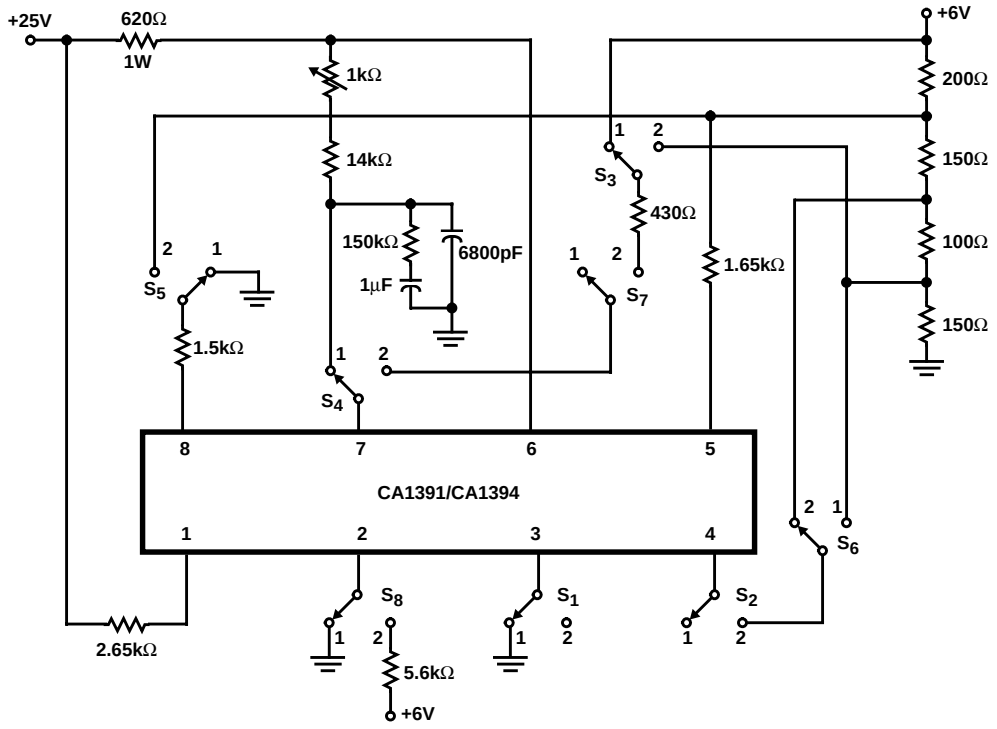
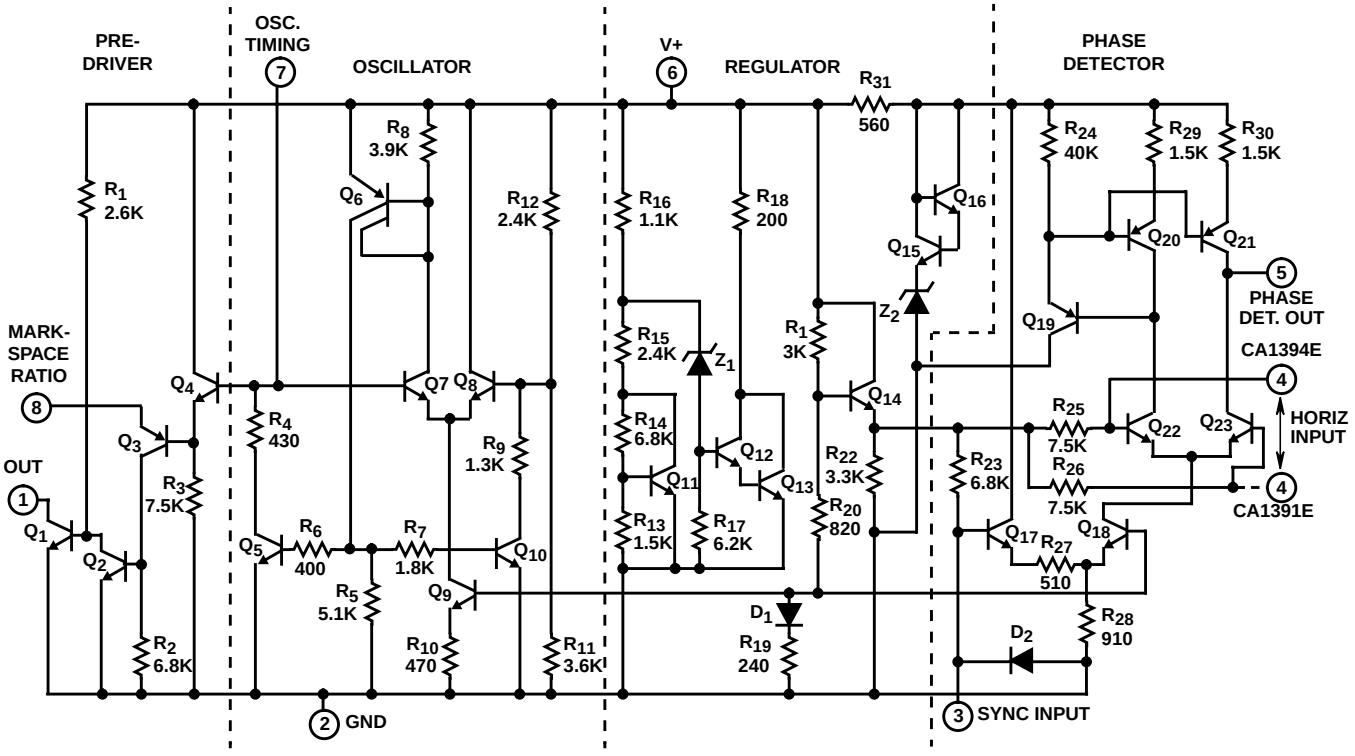


FIGURE 1. DC TEST CIRCUIT

Schematic Diagram



NOTE: All resistances are in ohms.

Application Information

Circuit Operation (See Schematic Diagram)

The CA1391 and CA1394 contain the oscillator, phase detector, and predriver sections necessary for the television horizontal oscillator and AFC loop.

The oscillator is an RC type with Terminal 7 used to control the timing. If it is assumed that Q_7 is initially off, then an external capacitor connected from Terminal 7 to ground charges through an external resistance connected between Terminals 6 and 7. As soon as the voltage at Terminal 7 exceeds the potential set at the base of Q_8 by resistors R_{11} and R_{12}, Q_7 turns on, and Q_6 supplies base current to Q_5 and Q_{10}. Transistor Q_5 discharges the capacitor through R_4 until the base bias of Q_7 falls below that of Q_8 at which time, Q_7 turns off, and the cycle repeats.

The sawtooth generated at the base of Q_4 appears across R_3 and turns off Q_3 whenever the sawtooth voltage rises to a value that exceeds the bias set at Terminal 8. By adjusting the potential at Terminal 8, the duty cycle at the pre-drive output (Terminal 1) may be changed. The phase detector is isolated from the remainder of the circuit by R_{31}, Z_2, Q_{15} and Q_{16}. The phase detector consists of the comparator Q_{22} and Q_{23}, and the gated current source Q_{18}. Negative going sync pulses at Terminal 3 turn off Q_{17}, and the current division between Q_{22} and Q_{23} is then determined by the phase relationship of the sync and the sawtooth waveform at Terminal 4, which is derived from the horizontal flyback pulse. If there is no phase difference between the sync and sawtooth, equal currents flow in the collectors of Q_{22} and Q_{23} during each half of the sync pulse period. The current in Q_{22} is turned around by current mirror Q_{20} and Q_{21} so that there is no net output current at Terminal 5 for balanced conditions. When a phase offset occurs, current flows either in or out of Terminal 5. In circuit applications, this terminal is connected to Terminal 7 through an external low pass filter, thereby controlling the oscillator.

Shunt regulation for the circuit is obtained by using a V_{BE} and zener multiplier. Resistors R_{13} and R_{14} multiply the V_{BE} of Q_{11}, and the ratio of R_{15} and R_{16} multiplies the voltage of the zener diode Z_1.

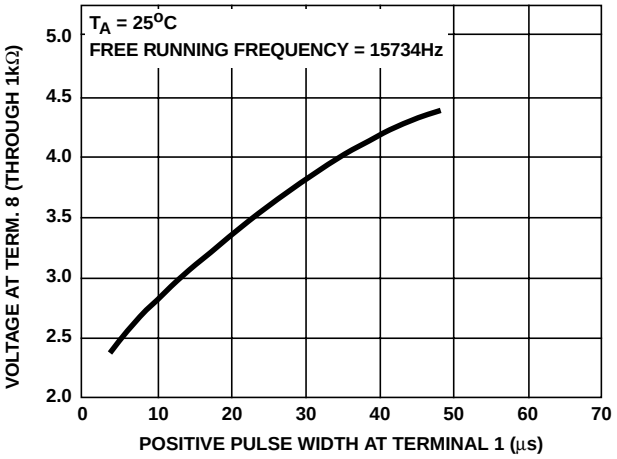


FIGURE 2. DUTY CYCLE AT THE PRE-DRIVE OUTPUT (TERMINAL 1) AS IT IS AFFECTED BY THE INPUT AT TERMINAL 8

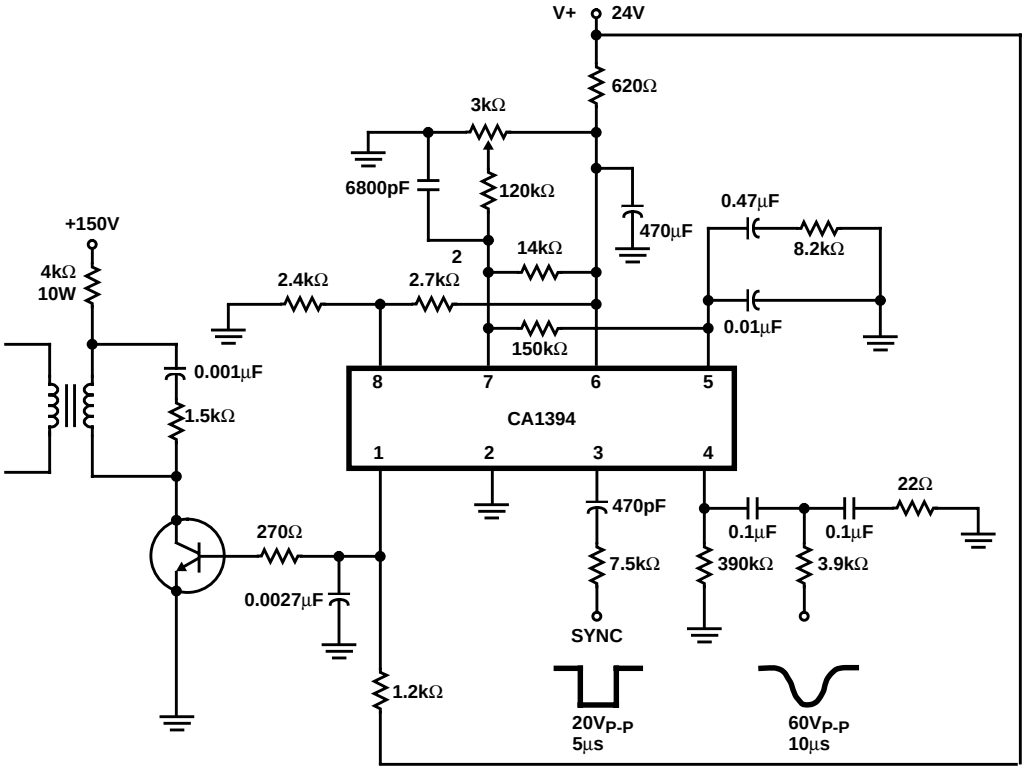


FIGURE 3. TYPICAL CIRCUIT APPLICATION